

(12) INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(19) World Intellectual Property Organization
International Bureau(43) International Publication Date
8 August 2002 (08.08.2002)

PCT

(10) International Publication Number
WO 02/061970 A1(51) International Patent Classification⁷: **H04B 7/08,**
H01Q 3/24(21) International Application Number: **PCT/GB02/00348**

(22) International Filing Date: 28 January 2002 (28.01.2002)

(25) Filing Language: English

(26) Publication Language: English

(30) Priority Data:
0102400.9 31 January 2001 (31.01.2001) GB
09/824,564 3 April 2001 (03.04.2001) US(71) Applicant (for all designated States except US): **QUINETIQ LIMITED** [GB/GB]; Registered Office, 85 Buckingham Gate, London SW1 6PD (GB).

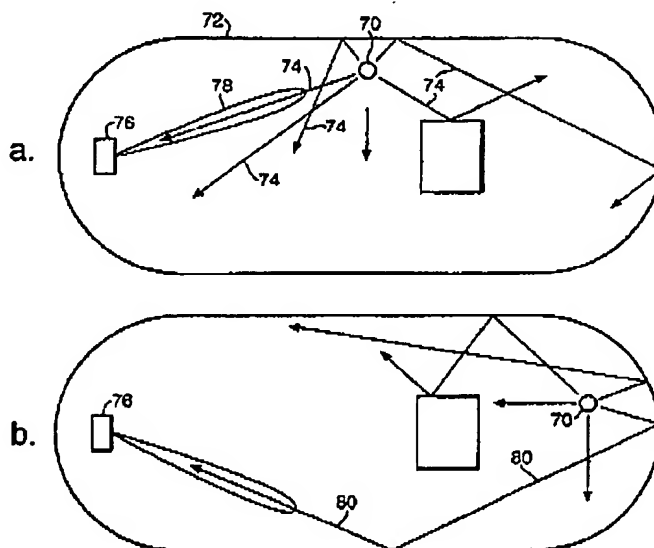
(72) Inventors; and

(75) Inventors/Applicants (for US only): **WATSON, Stephen** [GB/GB]; QinetiQ Malvern, St. Andrews Road, PE202, Malvern, Worcs WR14 3PS (GB). **DEAN, Michael**(GB/GB): QinetiQ Malvern, St. Andrews Road, PE306, Malvern, Worcs WR14 3PS (GB). **WALLACE, Neil** [GB/GB]: QinetiQ Malvern, St. Andrews Road, PE201, Malvern, Worcs WR14 3PS (GB). **SMITH, Andrew** [GB/GB]: QinetiQ Malvern, St. Andrews Road, PE306, Malvern, Worcs WR14 3PS (GB).(74) Agent: **DAVIES, P.**; D/TP QinetiQ Formalities, Cody Technology Park, A4 Building Ively Road, Farnborough, Hampshire GU14 0LX (GB).

(81) Designated States (national): AE, AG, AL, AM, AT, AU, AZ, BA, BB, BG, BR, BY, BZ, CA, CH, CN, CO, CR, CU, CZ, DE, DK, DM, DZ, EC, EE, ES, FI, GB, GD, GE, GL, GM, HR, HU, ID, IL, IN, IS, JP, KE, KG, KP, KR, KZ, LC, LK, LR, LS, LT, LU, LV, MA, MD, MG, MK, MN, MW, MX, MZ, NO, NZ, OM, PH, PL, PT, RO, RU, SD, SE, SG, SI, SK, SL, TJ, TM, TN, TR, TT, TZ, UA, UG, US, UZ, VN, YU, ZA, ZM, ZW.

(84) Designated States (regional): ARIPO patent (GH, GM, KE, LS, MW, MZ, SD, SL, SZ, TZ, UG, ZM, ZW), Eurasian patent (AM, AZ, BY, KG, KZ, MD, RU, TJ, TM),

[Continued on next page]

(54) Title: **SIGNAL DETECTION USING A PHASED ARRAY ANTENNA**

(57) Abstract: A method, and associated apparatus, for reading information from a transmitter (70) using a phased array antenna (76) wherein adjustment of the phased array antenna providing an RF link is performed in a period when the incident signal contains no information. In one embodiment, the method and apparatus are employed to receive and track an FM video signal transmitted by a mobile television radio-camera.

WO 02/061970 A1

WO 02/061970 A1

European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE, TR), OAPI patent (BF, BJ, CF, CG, CI, CM, GA, GN, GQ, GW, ML, MR, NE, SN, TD, TG).

Published:

— with international search report

Declaration under Rule 4.17:

— of inventorship (Rule 4.17(iv)) for US only

For two-letter codes and other abbreviations, refer to the "Guidance Notes on Codes and Abbreviations" appearing at the beginning of each regular issue of the PCT Gazette.

WO 02/061970

PCT/GB02/00348

SIGNAL DETECTION USING A PHASED ARRAY ANTENNA

This invention relates to a method and apparatus for receiving certain radio frequency (RF) signals using a phased array antenna, and in particular to a device and method for providing a RF link during any reconfiguration of the phased array antenna.

A typical wireless RF system uses a transmit antenna (transmitter) and a receive antenna (receiver) to support an RF link. In many environments, such as indoors or in enclosed outdoor spaces such as sports stadia, the transmitted signal will be reflected and may therefore reach the receiver via a multitude of different paths, each of different path length. These so called multipath effects can seriously degrade the quality of the received signal.

It is well established in the field of radio communications to use a directional receive antenna to reduce multipath effects. However, when a directional receive antenna is employed it must be constantly, and accurately, directed towards the transmitter.

An example of the use of a directional receive antenna is found in televising events. Radio cameras are used in Outside Broadcasts (OBs) to provide close in pictures of the event being televised. Currently most handheld radio cameras require a directional receiver that is rotated so that it is constantly aligned with the transmitter. This requirement is generally satisfied by using a standard dish antenna and a person (called a panner) who watches where the radio camera goes and rotates the directional receive antenna accordingly. The tracking must be precise and proves difficult if visibility is poor. In addition, if there is no direct line of sight between the receiver and the transmitter, the RF link is generally lost. Ensuring the receive antenna is pointed correctly becomes further complicated when the transmitter is moving.

The use of phased arrays, which are electronically controllable directional transmitter or receiver antenna, is well known in the art of RADAR technology. Traditional phased array antenna tend to comprise hundreds of elements, each with individual

WO 02/061970

PCT/GB02/00348

2

phase shifters working at the operating frequency of the antenna (often 1GHz and above). A description of such systems is given in N. Fourikis, 'Phased Array Based Systems And Applications', Wiley Interscience Publication, 1997, ISBN 0-471-01212-2.

WO97/03367 describes a phased array device that can be produced at a much lower cost, and is much smaller in size, than the traditional phased array systems. Instead of phase shifting the signals received at each antenna element at the operating frequency, the RF signals are down converted to a first and then to a second intermediate frequency. During the second down conversion, the phase of the second intermediate frequency signal is changed by controlling the phase of the corresponding local oscillator. As the phase shifting is performed at a much lower frequency than the RF signal, inexpensive devices are available that can provide a high level of phase control. This allows phase shifting to be performed with greater accuracy thereby enabling the number of elements in a phased array antenna to be reduced. These antennas are thus considerably cheaper to produce than the traditional phased array systems. Herein such devices are termed Low Cost (LC) phased array antenna.

To change the directional receive properties of a phased array antenna requires reconfiguration of the phased array by altering the phase and amplitude shifts applied to the signals received by each of the antenna elements. During any such period of phased array antenna reconfiguration there is a risk that the information being received by the phased array antenna will be corrupted.

It is an object of this invention to use a phased array antenna, in particular an LC phased array antenna, to acquire and track certain types of signal from an RF transmitter and to receive information from that transmitter without any substantial interruption to that receipt of information arising from phased array antenna reconfiguration.

WO 02/061970

PCT/GB02/00348

3

According to the first aspect of this invention a method of reading information from a transmitter comprises using a phased array antenna wherein adjustment of the phased array antenna providing an RF link is performed in a period when the incident signal contains no information.

Advantageously, any tracking of movement of the incident signal is also performed in one or more periods when the incident signal contains no information.

The present invention thus permits a phased array antenna to be adjusted, without an associated loss of information, whilst it is being used to receive information from a transmitter.

In a preferred embodiment the received signal is a frequency modulated (FM) analogue video signal, of a frequency between 12.2GHz and 12.5GHz, having an information carrying portion and a non-information carrying portion.

The use of the present invention for the reception of FM video signals ensures that there is no disturbance to the received video image during any alteration of the receive beam being used to provide the RF link. Furthermore, this invention allows any changes in the angle of incidence of signals reaching the phased array antenna to be tracked using the phased array antenna without any disturbance to the resultant video images.

Advantageously, the phased array antenna can simultaneously receive two or more incident signals thereby establishing two or more RF links. In a further preferred embodiment, a Low Cost (LC) phased array antenna is used.

Conveniently, a computer program, which may be held on an appropriate storage medium, can be used to implement the method described above.

WO 02/061970

PCT/GB02/00348

4

In a second aspect of this invention a control apparatus for a phase array antenna comprises input detection means for determining the direction of incidence of input signals and selecting therefrom a preferred input signal, means for adapting the phased array antenna to form a receive beam substantially in the direction of incidence of the preferred input signal, and tracking means to track any change in the direction of incidence of the preferred input signal and to adapt the phased array antenna accordingly to form a receive beam substantially in that direction, wherein adaptation of the phased array by the tracking means occurs substantially during periods in which the incident signal contains no information.

In a preferred embodiment, the tracking means tracks any change in the direction of incidence of the preferred input beam by comparing the signal strength received using the receive beam currently being used to receive the preferred input signal with the strength of any signal received by one or more other receive beams.

Advantageously, the control apparatus can also comprise two or more sets of beamforming hardware such that two or more receive beams may be simultaneously formed allowing two or more RF link to be established.

Preferably, the control apparatus further comprises means for extracting a signal, which may be an FM analogue video signal, from the phased array antenna.

The invention will now be described, by way of example only, with reference to the accompanying figures wherein;

Figure 1 shows the principle of operation of a phased array receiver;

Figure 2 illustrates the architecture of an LC phased array receiver for tracking an RF signal;

Figure 3 shows a schematic illustration of a typical analogue FM signal,

WO 02/061970

PCT/GB02/00348

5

Figure 4 illustrates the reception of signals, and

Figure 5 illustrates the use of a dual beam phased array receiver.

The operation of a general phased array antenna will now be described with reference to figure 1.

A phased array antenna receiver (1) comprises n antenna elements (2) which provide electrical signals (4) derived from an incident RF signal (not shown). Phase shifters (6) provide a phase shift to the electrical signals (4) producing phase shifted electrical signals (8). The phase shifted electrical signals (8) are then attenuated by an attenuation means (9) producing signals that are both phased shifted and attenuated (11). The signals that are both phased shifted and attenuated (11) are then combined by a combiner (10).

It is possible to make the phased array antenna receiver particularly sensitive to radiation incident from a certain direction. This is done by controlling both the phase shift applied to each of electrical signals (4), and the relative amplitude weighting given to each of the phased shifted electrical signals (8) by the attenuation means (9). As described later in detail, there are several techniques of applying phase shifts to the electrical signals (4).

Selecting phase shifts and amplitude weightings that cause the phased array antenna receiver (1) to have directionally dependent RF signal reception properties is termed beam forming. For example, figure 1 shows a beam (12) that could be formed by applying certain phase shifts and amplitude weightings to the electrical signals (4) that are produced by the n antenna elements (2). Alternatively, different phase shifts and amplitude weightings could be applied to produce another beam (14).

WO 02/061970

PCT/GB02/00348

6

A receive beam, also simply termed a beam, is the angular range over which the detector is sensitive to incident signals. In other words, a receive beam can be considered as a three dimensional area in space and the phased array antenna will be sensitive to any signal incident on it from that three dimensional area. In reality, it is unlikely that perfect beams would be formed; each receive beam would have associated "sidelobes". Methods of beamforming and the existence and suppression of sidelobes (as described for LC systems with reference to figure 2) are well known to persons skilled in the art of phased array RADAR.

The principle of operation of an LC phased array antenna will now be described with reference to figure 2.

The traditional design philosophy for phased array RADAR systems has been that each element uses an individual phase shifter for phase control. The phase shifter is typically a Monolithic Microwave Integrated Circuit (MMIC) and is characterised by a high cost due to limited production runs and the fact that the device has to function at the operating frequency of the antenna (often 1GHz and above). The phase shifter is controlled by a digital input. Standard devices are 4 bit giving 22.5° of phase resolution, whilst more complex options have 6 bits that provide approximately 6° of phase resolution. As a result of this relatively low level of phase control, in order for the antenna to be able to scan the beam in 1° or sub-degree steps, hundreds or thousands of elements are required. Hence traditional phased arrays have used hundreds or thousands of expensive MMIC phase shifters and consequently have been utilised almost exclusively by the military for large installations.

It is possible to avoid using individual phase shifters for phase control without the need to employ expensive digital beamforming solutions using the beamforming architecture disclosed in WO97/03367.

An LC phased array receiver comprises a plurality of antenna elements (22a, b, c). The electrical signal produced by each antenna element when receiving an RF signal is

WO 02/061970

PCT/GB02/00348

7

amplified by low noise amplifiers (24), passes through image reject filters (26) before being down-converted to a first intermediate frequency signal (32) by means of microwave mixers (28). A microwave local oscillator signal (30) is used by the microwave mixers (28) in the down conversion process. The first intermediate frequency signals (32) are then fed into the beamforming hardware (21).

On entering the beamforming hardware (21) the first intermediate frequency signals (32) passes through amplifiers (34), and image reject filters (36), before being down-converted to second intermediate frequency signals (46) by intermediate frequency mixers (38). During this second down-conversion process, phase shifts are introduced by changing the phase of the second local oscillator (LO) signals (44a,b,c) using phase shifter means (42) and are then applied to each of the intermediate frequency mixers (38). The phase shift introduced by the phase shifter means (42) is controlled by a digital control bus (52), and produces second intermediate frequency phase shifted electrical signals (50a,b,c).

Because the phase shifter means (42) used to phase shift the second LO signal operates at a frequency much lower than the RF signal, inexpensive vector modulator devices can be used. A person skilled in the art would be aware of the various types of vector modulator device that would be suitable for this purpose. Typical vector modulator devices, such as those used in mobile phones, are controlled by low cost 12 bit digital-to-analogue converters and provide a very high level (sub 1°) of phase control.

The second intermediate frequency phase shifted electrical signals (50a,b,c) are combined in the combiner (54). A suitable frequency for the second IF electrical signal is 70MHz. After being combined, the resultant signal (55) is split two ways. One part is cabled into a power detect module (56) whilst the other passes through an Automatic Gain Control (AGC) module (58). After the AGC module, the signal is again split two ways. One part is routed as the output of the antenna (62), whilst the other passes through a suitable demodulator or decoder module (59).

WO 02/061970

PCT/GB02/00348

8

The output of the power detect module (56) can be used by the microcontroller (60) to determine the best position to point the receive beam. The microcontroller (60) also controls, over the digital bus (52), the phase shift that is applied to each second local oscillator signal (44a,b,c) by the phase shifter means (42). The AGC module (58) works to keep the output signal (62) at a constant power level of +5dBm without compromising the linearity of the receive chain. The power detect module (56) and the AGC module (58) work independently of the signal's modulation, and as a result the phased array antenna can acquire and track analogue or digital signals.

The decoder module (59) demodulates or decodes part of the output signal (62) into baseband components. In the case of FM video, the various components of the video can then be measured and may be used to assess the quality of the video signal that is being received. It is thus possible for the microcontroller (60) to use a video signal quality measurement from the decoder module (59) instead of, or as well as, the signal strength measurements provided by the power detect module (56) when deciding how to direct the receive beams.

In a device operating in accordance to this invention, a synchronisation signal is provided by the decoder module (59) to the microcontroller (60) to indicate when the received signal contains no information. The microcontroller (60) only reconfigures the phased array antenna during these periods; hereinafter termed the non-information carrying period.

An example of a signal having a non-information carrying period will now be described with reference to figure 3.

A FM analogue video signal of a given period (63), typically 20ms, comprises an information carrying period (64) of typically 18.5ms and a non-information carrying period (65) of approximately 1.5ms. The non-information carrying period (65) is commonly termed the "fly-back" portion of the signal. All the video information is

WO 02/061970

PCT/GB02/00348

9

contained in the information carrying period (64), and there will be no perceivable interference to the displayed video image if reconfiguration of the phased array antenna is performed during the non-information carrying period (65).

This technique can be applied to any signal, analogue or digital, having a non-information carrying period. For example, a digital signal could be transmitted that contains information for a certain period but is configured to have a non-information carrying period. A person skilled in the art could produce appropriate data buffering systems to ensure continuity of the digital output of data.

The output signal (62) can be routed from the phased array to an Antenna Control Unit (not shown) via a standard tri-axial cable. This cable can also be used to support the control and telemetry data between the phased array and the Antenna Control Unit (ACU) and provide a power supply for the array. The ACU can be located at a convenient position, which may be remote to the phased array antenna itself.

In this embodiment the ACU's function is to provide a suitable interface which the operator can use to control the phased array. However, a person skilled in the art would recognise that many different methods of routing the received signal and control data could be employed (e.g. fibre optic, low frequency radio data links). In addition, the ACU can be fitted with a decoder or demodulator as specified by the user. These options do not affect the fundamental principles underlying this invention and are merely workshop variations which would be immediately apparent to a person skilled in the art.

In order to obtain a discrete set of beams from an LC phased array, the phased array antenna is calibrated before use. A discrete set of beams (for example $+50^\circ$ to -50° in 1° steps) can be calibrated for a given operating frequency or for groups of frequencies within a given band. For each beam there is a phase and amplitude weighting for each element of the antenna. The calibration data is stored, and subsequently used to allow the formation of a given directional beam for a given frequency. During calibration

WO 02/061970

PCT/GB02/00348

10

the absolute phase between each of the calibrated beams can be controlled so that it is the same value for each beam. This helps to minimise phase interference whilst switching beams.

In addition to performing a calibration at each operating frequency it is also possible to perform several different calibration types. One set implements zero amplitude attenuation on each element. This provides maximum gain in the main beam, but the sidelobe levels are not controlled. Conversely, a fully weighted calibration set provides maximum sidelobe suppression which results in a reduction in the receiver's susceptibility to multipath effects. The disadvantage of a fully weighted calibration is that the algorithms used to synthesise such beams tend to reduce the gain of the antenna. In addition to the two calibration types described here, there are a multiplicity of calibration options that can be used for a variety of beam patterns. Such calibration types are well known to those skilled in the art of phased array RADAR technology.

A result of the high level of phase control provided by the LC system described above is that beams can be synthesised and scanned in sub-degree steps from arrays of very few elements (for example 8 or 16 elements). It is also possible to have modular RF front end and beamforming circuits. A typical module consists of 8 radiating elements complete with superheterodyne receiver and phase control circuit. The modules can also be grouped together so as to create a linear or planar phased array antenna that satisfies the system requirements.

For example, two 8 element modules have been combined to produce a 16 element linear phased array antenna. More modules could be combined, for example if a more directional antenna were required. A larger array would have more gain that could support an RF link from a given transmitter over a longer distance. The 16 element array will support an RF link with a conventional handheld radio camera over distances of up to 1km.

WO 02/061970

PCT/GB02/00348

11

The acquisition arc (i.e. lateral angular range over which beams can be formed) for a 16 element phased array is approximately 100°. Scanning beyond $\pm 50^\circ$ is possible but at the expense of some degradation in the beam pattern such as increased sidelobe levels and a broadening of the main beam. Supporting an RF link over larger angles is achievable in several ways. A combination of receivers can be located so as the transmitter is always within the acquisition arc of the network, with handovers between arrays occurring automatically at the various boundaries. Alternatively a single receiver can be mounted onto a turntable and the servo driven by control signals generated by the array. A third option is to use a curved RF front end instead of a linear row. Curved surface and full circular arrays have been developed that provide 360° of coverage.

A further advantage of LC phased array devices over traditional phased arrays is that they work independently of the operating frequency of the antenna. Because the beamforming is performed at a low intermediate frequency, the RF frequency of the antenna is unrestricted. Whatever the operating frequency, the RF signal is downconverted to the necessary IF and the phase control implemented using the second IF mixer. This type of detector is thus totally 'modular' in frequency; it can be used to receive an RF signal of any frequency.

For FM video link applications, frequencies within the 2GHz or 12GHz radio camera bands are generally used. For example, 12 25MHz channels could be provided between 12.2125GHz and 12.4875GHz. An LC phased array device can thus be built which can track a radio camera transmitting at an allocated 12GHz frequency channel with an output of 70MHz, +5dBm (the industry standard).

It is also possible to include additional sets of beamforming hardware in phased array devices. Simultaneous formation of a plurality of receive beams is well known to a person skilled in the art of phased array RADAR. To simultaneously form multiple receive beams using an LC device, the first intermediate frequency signals (32) are divided and supplied to a plurality of sets of beamforming hardware (21). Each set of

WO 02/061970

PCT/GB02/00348

12

beamforming hardware produces output signals from its power detect, AGC, and decoder modules. A single microcontroller can then be used to direct the receive beams associated with each set of beamforming hardware.

The use of a phased array receiver to acquire and track a transmitted RF signal will now be described, with reference to figure 4. Although the LC phased array receiver described with reference to figure 2 is particularly suitable for implementing the transmitter tracking methods described below, a person skilled in the art would recognise that any phased array receiver could be employed.

When a signal is transmitted by an omni-directional transmitter (70) in an enclosed environment, such as a sports stadium (72), a plurality of multipath RF signals (74a, 74b, 74c, 74d, 74e) are produced. If an omni-directional receiver were used to receive the transmitted signal the many multi-path signals, all of which are slightly out of phase due to travelling along paths of different length, would all be received producing a resultant received signal that has a high level of multi-path interference.

As shown in figure 4a and as described above, a phased array receiver (76) can be used to form a directional receive beam (78) which reduces susceptibility to multi-path interference effects. Figure 4b shows the use of a phased array receiver (76) to receive a reflected signal (80) from an omni-directional transmitter (70) in the absence of any direct line of sight path.

The phased array receiver system must initially ascertain the angle of incidence of a suitable RF signal. This is generally performed by determining the direction from which the strongest transmitted signal originates. Alternatively, the angle of incidence that provides a signal of acceptable strength with the lowest level of multi-path interference (i.e. provides the highest quality signal) could be selected. The strongest, or highest quality, signal may be the line of sight signal, but it may also be a reflection. The process of determining the angle of incidence of a suitable RF signal is herein termed an acquisition scan.

WO 02/061970

PCT/GB02/00348

13

For a full acquisition scan, a typical LC phased array of the type described with reference to figure 2 can sequentially load a full set of beams from the selected calibration set (e.g. from $+50^\circ$ to -50° in 1° steps). For each beam loaded, the power of the received signal is measured and the beam that gave the highest reading is selected as the centre beam for a 'mini-scan'. A mini-scan is the same as a full scan but over a much narrower range, and possibly of a higher angular resolution.

The operator can control the angular range over which the initial scan takes place and the number of degrees between each step (1° , 2° etc), or alternatively can chose to load a single fixed beam. Using a typical LC phased array of the type described with reference to figure 2, acquisition of a signal over a 100° arc takes approximately 0.4 seconds. Faster rates can be achieved by using a faster processor.

The result of the acquisition scan determines the angle of incidence of the preferred RF signal. Once a preferred signal has been acquired, any change in the angle of incidence of the signal on the phased array receiver can be tracked. The initiation of a tracking routine can be controlled manually, or automatically executed at the end of the acquisition scan. The tracking routine allows for any movement of the transmitter, phased array receiver or intervening objects.

A person skilled in the art would recognise that there are several tracking routines that may be used. An example of a tracking routine is free running dither. In this routine the array loads a beam first to the left of the current position, and then to the right. The received signal power of the two dithered beams is measured and the results compared with that of the current centre beam. The beam that gives the highest value then becomes the centre beam for the next dither routine. However, unless the tracking steps are performed during a non-information carrying period of the signal some of the information contained in the signal will be lost.

In accordance with this invention, a controlled dither technique can be used to minimise data loss during the tracking process. In the case of analogue FM video

WO 02/061970

PCT/GB02/00348

14

signals, beams are only loaded during the non-information carrying period of the signal. In other words, reconfiguration of the phased array is performed only when the microcontroller (60) receives a frame synchronisation pulse from the decoder module (59). This ensures that the picture interval of the frame is undisturbed and minimises visible picture interference.

The process of tracking a signal obviously requires more than one reconfiguration of the phased array antenna. The speed of reconfiguration of the phased array is determined by the speed of the microcontroller (60) and the associated electronics. Different types of signal will also have different non-information carrying periods of time.

For certain signal types and phased array systems it may be possible to perform sufficient reconfigurations of the phased array antenna during the non-information carrying period to perform a tracking routine which loads a beam first to the left of the current position, and then to the right of that position and selects which beam is to be used to receive during the next period; i.e. perform a left/right tracking procedure. This would be preferable if the variation of the angle of incidence of the signal on the phased array antenna was changing rapidly with time.

For a typical 66MHz microcontroller, reconfiguration of the phased array takes approximately 0.8ms. Following reconfiguration, it takes approximately 3.2ms to obtain a measure of signal strength or quality. The signal strength or quality measurements can however be performed during the information carrying period of the signal without any detrimental effect on the receipt of information.

An FM video signal typically has a 1.5ms non-information carrying period. A beam to the left of the current position may thus be loaded during one non-information carrying period and then a beam to the right of the current position loaded during the subsequent non-information carrying period. In this way movement in the angle of incidence of signals may be tracked.

WO 02/061970

PCT/GB02/00348

15

If a signal had a longer non-information carrying period, or the speed of the microcontroller was increased, it would be possible to perform the left/right tracking procedure, with associated measurement of signal strength or quality, during the non-information carrying period.

According to the environment the number of beams that make up a tracking routine, as well as the angular separation between each beam, can be varied. The use of an increased number of beams during the tracking procedure will increase the time required for the tracking process, and may require a single tracking step to be performed over more than one non-information carrying period.

An acquisition scan can also be activated periodically, if the signal strength drops below a certain threshold or manually by an operator of the system.

When tracking an RF transmitter, the phased array receiver will generally use the line of sight path to support the RF link. If the line of sight link is lost, the phased array receiver can automatically start to scan the acquisition arc and locate any reflected signals being produced as a result of the operating environment. The strongest, or highest quality, reflected signal can then be acquired and tracked until the line of sight path becomes available again. In this way an RF link can be supported, even when the RF transmitter is not line of sight.

In addition, there may also be some situations when the line of sight path may not provide the best RF link performance; for example when both multi-path signals and the line of sight signal are incident on the phased array receiver within the receive beam. In this case the phased array antenna can select not to acquire the line of sight signal, but instead acquire and track a reflected signal that provides a higher quality video image.

WO 02/061970

PCT/GB02/00348

16

It should be noted that the system can also be used if the transmitter and receiver are in fixed positions, but objects move into the direct line of sight or if objects from which the signal is being reflected change position.

As described above with reference to figure 2, a plurality of independent receive beams may be formed using an LC phased array device. The use of multiple receive beams to acquire and track a transmitted signal will now be described with reference to figure 5.

Figure 5a shows a single phased array receiver (100), simultaneously forming a first receive beam (102) and a second receive beam (104). The first receive beam (102) and the second receive beam (104) can independently acquire and track a first transmitter (106) and a second transmitter (108). The first transmitter (106) and the second transmitter (108) must be transmitting at different frequencies.

In this configuration, the beams are independently steered and hence support links to transmitters operating at different frequencies within the system bandwidth. In this way a single antenna array, with a plurality of beam-forming hardware, could be used to track a plurality of transmitters.

Alternatively, two independent beams operating at the same frequency can be used to improve tracking. Figure 5b shows a single phased array receiver (100), forming a first receive beam (112) that acquires and tracks one signal (114) transmitted by transmitter (110). A second receive beam (116) then sequentially forms beams across the acquisition arc searching for the optimum receive beam direction for the first receive beam (112) to adopt. Once an optimum receive direction has been established by the second receive beam, the first receive beam is directed accordingly. To ensure disruption to the RF link is minimised, this redirection of the first receive beam is performed during non-information carrying periods of the signal.

WO 02/061970

PCT/GB02/00348

17

The example given in figure 5b refers to two independent beams, but this should not be seen as limiting. One or more beams can be dedicated to supporting RF links, whilst one or more additional beams can be continually scanning the acquisition arc searching for the beam position that will provide the best link performances for the next time slot. To ensure disruption to the RF link is minimised, any redirection of the beams providing an RF link is undertaken during non-information carrying periods of the signal.

In addition to the use of independent beams formed from a single phase centre as described above, the phased array antenna can be configured so as to produce two or more beams from separate phase centres; this is called beam diversity and is well known to those skilled in the art of phased array RADAR. The beam diversity is obtained by using a subset of the array elements of the antenna to form beams. For example, if a 16 element LC array were used, two sets of 8 elements could be used so as to form beams from two diverse phase centres.

Figure 5c shows how two independent beams (120 and 122) can be formed from two phase centres (124 and 126) on an LC phased array antenna. Each set of beams originating from a phase centre can be controlled independently of the other beam sets. The beams can be controlled to track a single, or multiple, transmitters in the same way as beams originating from a single phase centre as described with reference to figures 5b and 5a. Again, disruption to the RF link is minimised by redirecting the beams supporting RF links during non-information carrying periods of the signal.

If, as shown in figure 5c, the two beams (120 and 122) both acquire and track a single transmitter (128) two separate links with the transmitter are provided. In this case, each link with the transmitter will be susceptible to different multipath interference effects because of the different position of each phase centre. The beam providing the signal output of the highest quality can thus be selected to provide the RF link. The use of diverse beams can hence be used to provide greater resistance to multi-path effects.

WO 02/061970

PCT/GB02/00348

18

Claims

- 1 A method of reading information from a transmitter using a phased array antenna wherein adjustment of the phased array antenna providing an RF link is performed in a period when the incident signal contains no information.
- 2 A method of reading information from a transmitter as claimed in claim 1 wherein tracking of any movement of the incident signal is performed in one or more periods when the incident signal contains no information.
- 3 A method of reading information from a transmitter according to any preceding claim wherein the received signal is a frequency modulated (FM) analogue video signal having an information carrying and non-information carrying portion.
- 4 A method of reading information from a transmitter according to claim 3 wherein the frequency of the transmitted video signal is between 12.2GHz and 12.5GHz.
- 5 A method of reading information from a transmitter according to any preceding claim wherein the phased array antenna can simultaneously receive two or more incident signals thereby establishing two or more RF links.
- 6 A method of reading information from a transmitter according to any preceding claim wherein a Low Cost (LC) phased array antenna is used.
- 7 A method of reading information from a transmitter as substantially hereinbefore described with reference to figures 2-5.
- 8 A computer program for controlling a phased array receiver incorporating the method of any preceding claim.

WO 02/061970

PCT/GB02/00348

19

9 A storage medium holding a computer program according to claim 8.

10 A control apparatus for a phased array antenna configured so as to implement the method claimed in claims 1 to 7.

11 A control apparatus for a phase array antenna comprising;

input detection means for determining the direction of incidence of input signals and selecting therefrom a preferred input signal;

means for adapting the phased array antenna to form a receive beam substantially in the direction of incidence of the preferred input signal; and

tracking means to track any change in the direction of incidence of the preferred input signal and to adapt the phased array antenna accordingly to form a receive beam substantially in that direction,

wherein adaptation of the phased array by the tracking means occurs substantially during periods in which the incident signal contains no information.

12 A control apparatus for a phased array antenna as claimed in claim 11 wherein the tracking means tracks any change in the direction of incidence of the preferred input beam by comparing the signal strength received using the receive beam currently being used to receive the preferred input signal with the strength of any signal received by one or more other receive beams.

13 A control apparatus for a phased array antenna as claimed in claim 11 or 12 wherein the phased array antenna comprises two or more sets of beamforming hardware such that two or more receive beams may be simultaneously formed allowing two or more RF links to be established.

WO 02/061970

PCT/GB02/00348

20

14 A control apparatus for a phased array antenna as claimed in any of claims 11-13 and further comprising means for extracting a signal from the phased array antenna.

15 A control apparatus for a phased array antenna as claimed in any of claims 11-14 wherein the input signal is a FM analogue video signal.

16 A control apparatus for a phased array antenna as substantially hereinbefore described with reference to figures 2-5.

WO 02/061970

PCT/GB02/00348

1/5

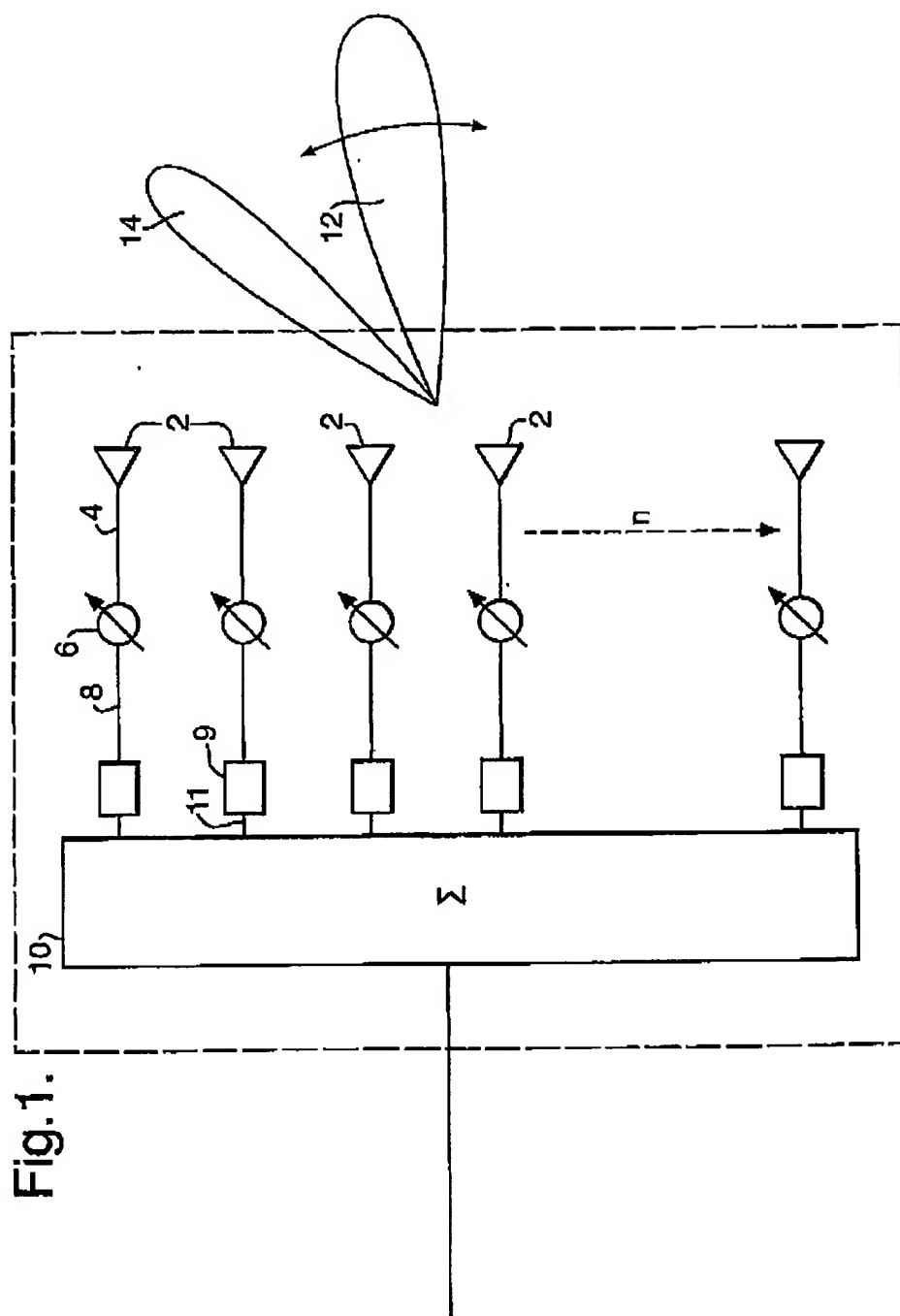


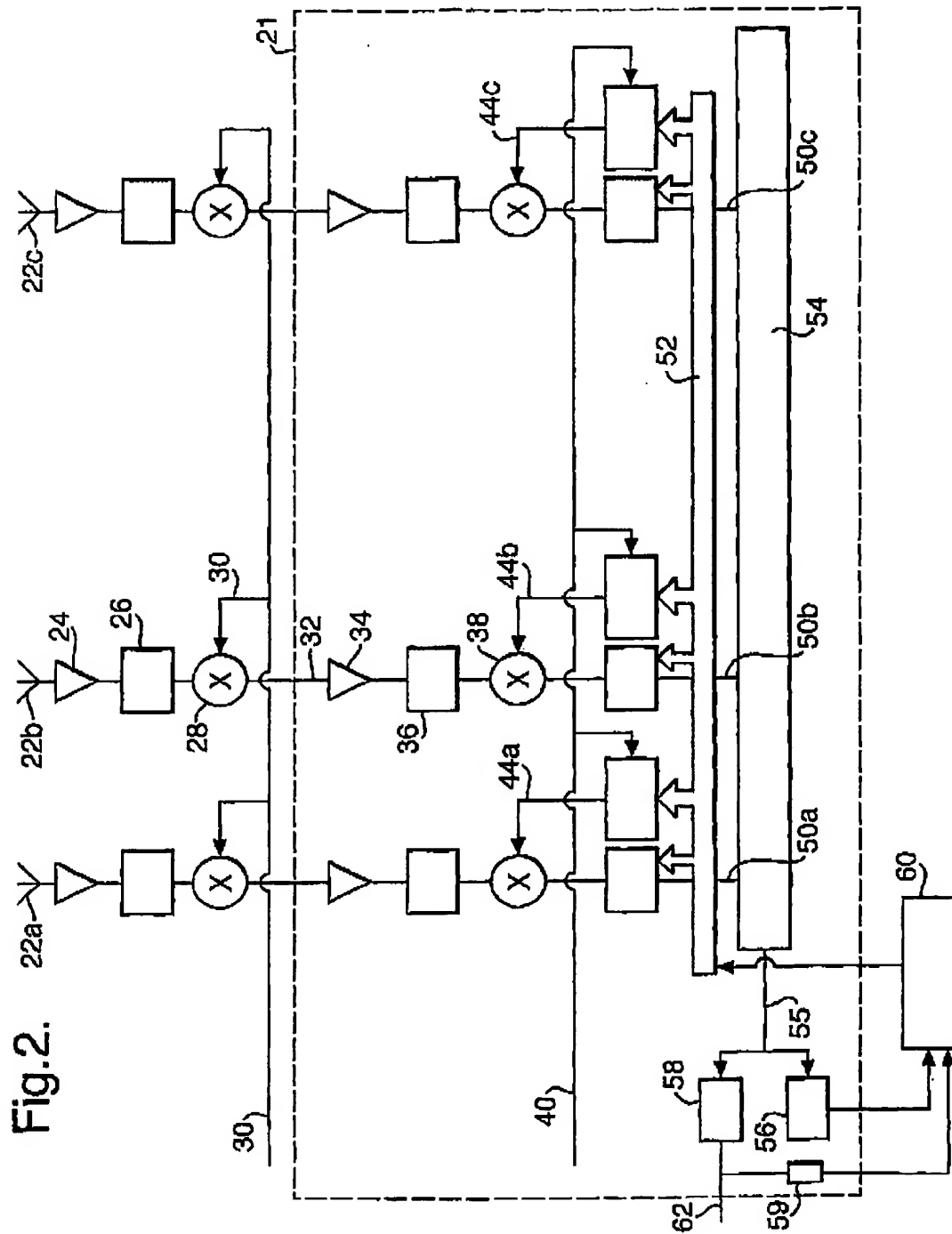
Fig. 1.

SUBSTITUTE SHEET (RULE 26)

WO 02/061970

PCT/GB02/00348

2/5



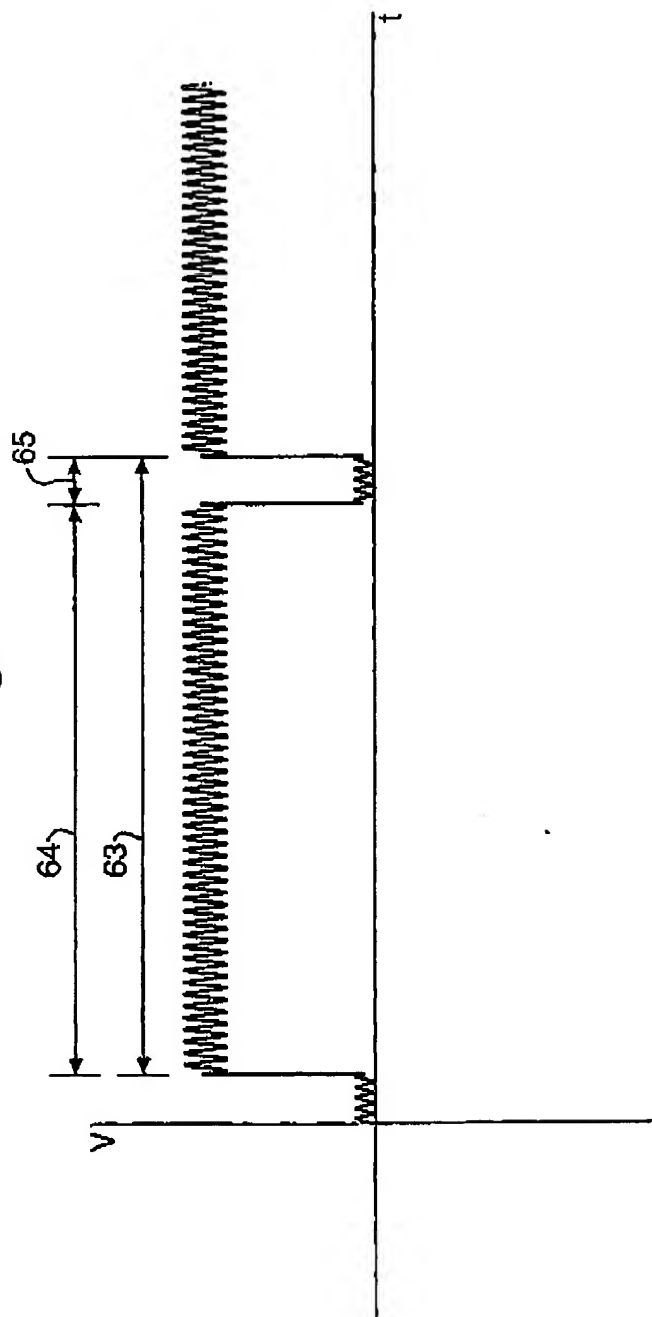
SUBSTITUTE SHEET (RULE 26)

WO 02/061970

PCT/GB02/00348

3/5

Fig.3.



SUBSTITUTE SHEET (RULE 26)

WO 02/061970

PCT/GB02/00348

4/5

Fig.4a.

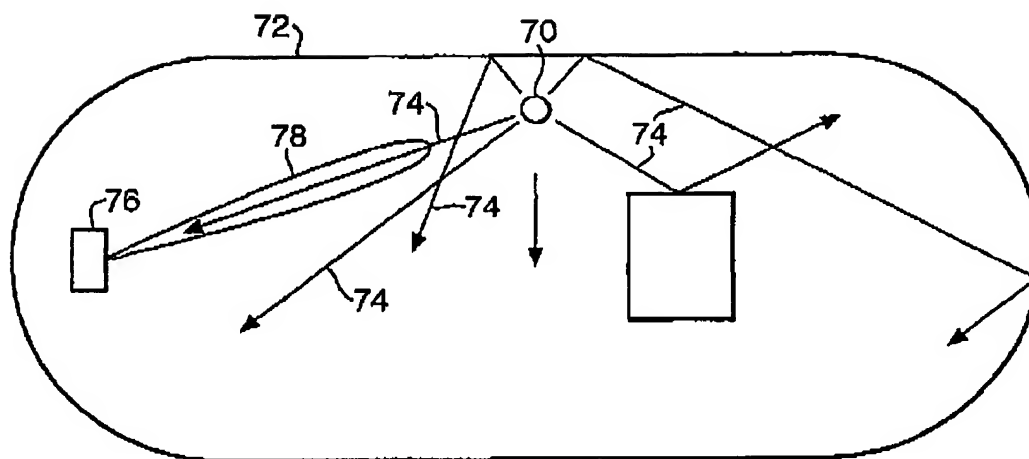
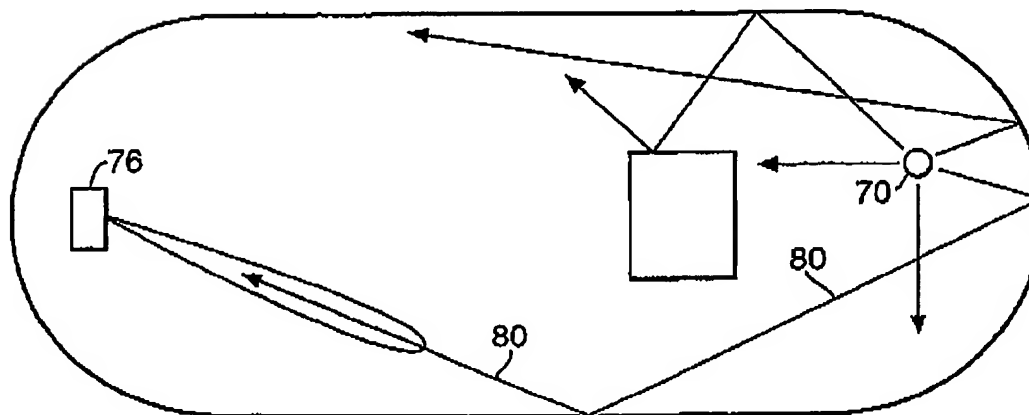


Fig.4b.



SUBSTITUTE SHEET (RULE 26)

WO 02/061270

PCT/GB02/00348

5/5

Fig.5a.

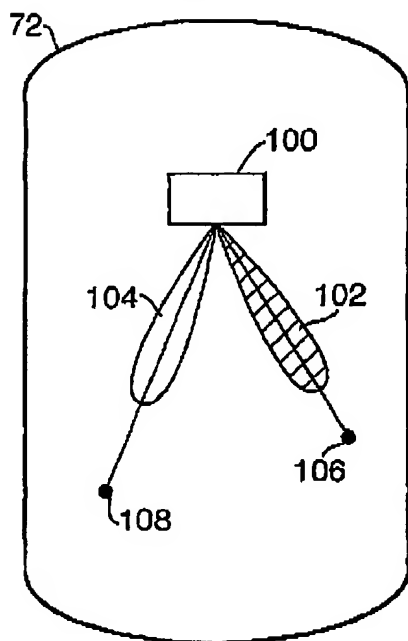


Fig.5b.

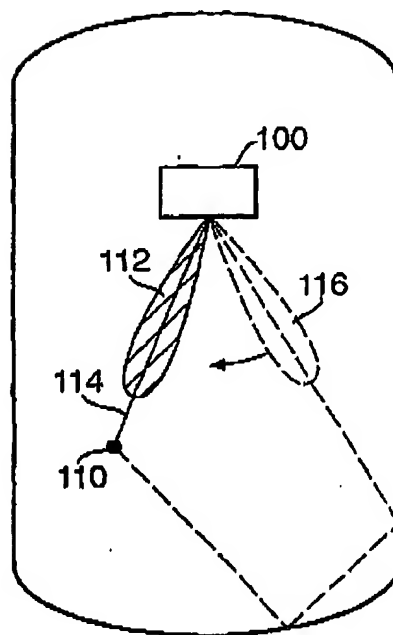
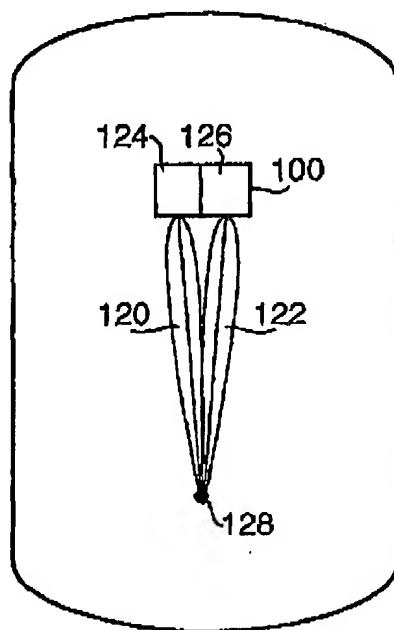


Fig.5c.



SUBSTITUTE SHEET (RULE 26)

SEP 15 2006

INTERNATIONAL SEARCH REPORT

International Application No
PCT/GB 02/00348A. CLASSIFICATION OF SUBJECT MATTER
IPC 7 H04B7/08 H01Q3/24

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
IPC 7 H04B H01Q H04N

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data, PAJ

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
Y	GB 2 196 211 A (BRITISH BROADCASTING CORP) 20 April 1988 (1988-04-20) abstract; figure 1 page 1, line 85 - line 96 page 5, line 23 - line 33 page 5, line 51 - line 67	1-6, 8-15
Y	GB 2 313 261 A (MOTOROLA LTD) 19 November 1997 (1997-11-19) page 5, line 1 - line 22 page 6, line 5 - line 18 page 9, line 9 - page 12, line 14 page 22, line 17 - page 24, line 5 page 29, paragraph 3 figure 1	1-6, 8-15

☒ Further documents are listed in the continuation of box C.☒ Patent family members are listed in annex.

* Special categories of cited documents:

- *A* document defining the general state of the art which is not considered to be of particular relevance
- *E* earlier document but published on or after the international filing date
- *L* document which may throw doubt on priority claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- *O* document relating to an oral disclosure, use, exhibition or other means
- *P* document published prior to the international filing date but later than the priority date claimed

- *T* later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
- *X* document of particular relevance: the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- *Y* document of particular relevance: the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art
- *Z* document member of the same patent family

Date of the actual completion of the international search

26 Apr11 2002

Date of mailing of the international search report

16/05/2002

Name and mailing address of the ISA

European Patent Office, P.O. Box 5918 Patentkan 2
NL - 2280 HV Rijswijk
Tel: (+31-70) 340-2040, Tx: 31 651 epo nl
Fax: (+31-70) 340-0010

Authorized officer

Bodin, C-M

Form PCT/ISA/210 (second sheet) (July 1992)

page 1 of 2

INTERNATIONAL SEARCH REPORT

International Application No

PCT/GB 02/00348

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	PATENT ABSTRACTS OF JAPAN vol. 013, no. 477 (M-885), 27 October 1989 (1989-10-27) & JP 01 188748 A (NISSAN MOTOR CO LTD), 28 July 1989 (1989-07-28) abstract	1-4
A	DE 40 06 295 A (KRAUSS MAFFEI AG) 30 August 1990 (1990-08-30) column 1, line 32 - line 37 abstract figures 1,2	1,2
A	US 5 303 240 A (BORRAS JAINE A ET AL) 12 April 1994 (1994-04-12) column 4, line 32 - line 34 column 4, line 51 - line 62	1,2

Form PCT/ISA/210 (continuation of second sheet) (July 1992)

page 2 of 2

SEP 15 2006

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/GB 02/00348

Patent document cited in search report		Publication date	Patent family member(s)	Publication date
GB 2196211	A	20-04-1988	NONE	
GB 2313261	A	19-11-1997	AU 723992 B2	07-09-2000
			AU 1782897 A	20-11-1997
			BR 9703357 A	15-09-1998
			CA 2202829 A1	17-11-1997
			CN 1170282 A	14-01-1998
			DE 69705356 D1	02-08-2001
			EP 0807989 A1	19-11-1997
			HU 9700908 A2	28-04-1998
			JP 10117162 A	06-05-1998
			RU 2141168 C1	10-11-1999
			US 6192256 B1	20-02-2001
			US 5999826 A	07-12-1999
JP 01188748	A	28-07-1989	NONE	
DE 4006295	A	30-08-1990	DE 4006295 A1	30-08-1990
US 5303240	A	12-04-1994	AU 2310092 A	11-02-1993
			CA 2110712 A1	21-01-1993
			EP 0636289 A1	01-02-1995
			WO 9301663 A1	21-01-1993

Form PCT/IBA/210 (patent family annex) (July 1992)